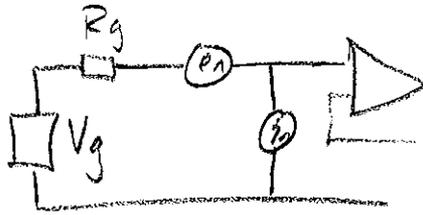


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1 a)



$$V_n = \sqrt{e_n^2 + (R_g i_n)^2}$$

Uncorrelated sources

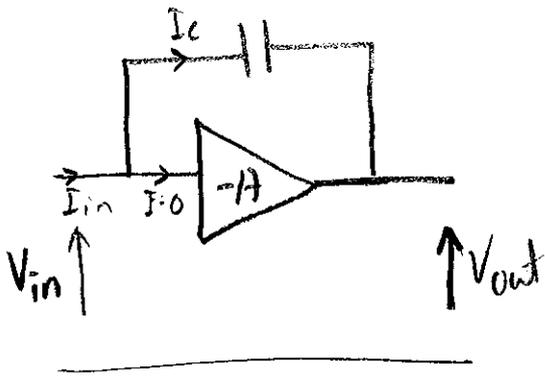
→ The noise power contribution should be added

b] A current through the device can be controlled by an applied voltage/current in a FET/bipolar transistor

c] The feed-back is increasing the linearity and reducing the gain.

d] A switched mode power supply is in general working at higher frequencies than 50 Hz and less energy need to be stored in each cycle enabling smaller transformers to be used.

2]

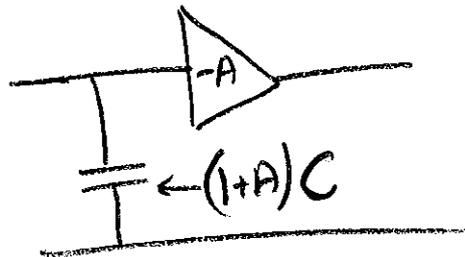


$$V_{out} = -A \cdot V_{in}$$

$$I_{in} = I_c = \frac{V_{in} - V_{out}}{j\omega C} = \frac{V_{in} - (-A \cdot V_{in})}{j\omega C}$$

$$= V_{in} (1+A) j\omega C$$

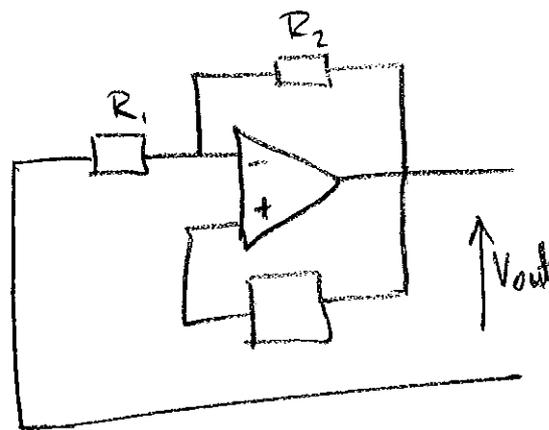
$$Z_{in} = \frac{V_{in}}{I_{in}} = \frac{V_{in}}{V_{in} (1+A) j\omega C} = \frac{1}{j\omega (1+A) C}$$



3] a) The overall gain should be 1 and the phase shift should be 0.

b) The lead-lag filter gives gain $1/3$ with a phase shift $\phi=0$. (Frequency axis in phase diagram is accidentally shifted)
 The lead-lag filter is connected to the positive input. The resistor network should thus compensate and give amplification $A_{R_1, R_2} = 3$.

c]



$R_1 - R_2$ forms a non inverting amplifier

$$A_v = \frac{R_1 + R_2}{R_1}$$

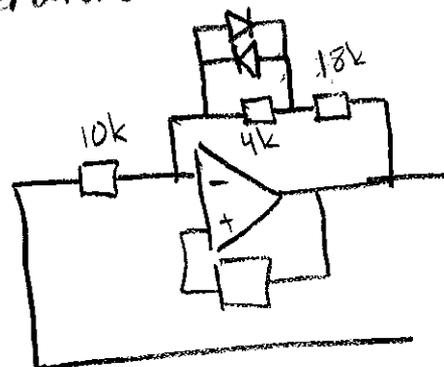
$R_2 = 2 \cdot R_1$ fulfills the theoretical conditions for oscillation let $R_1 = 10 \text{ k}\Omega$ $R_2 = 20 \text{ k}\Omega$

However resistors are not exact and parameters change with temperature so it does not give a stable amplitude

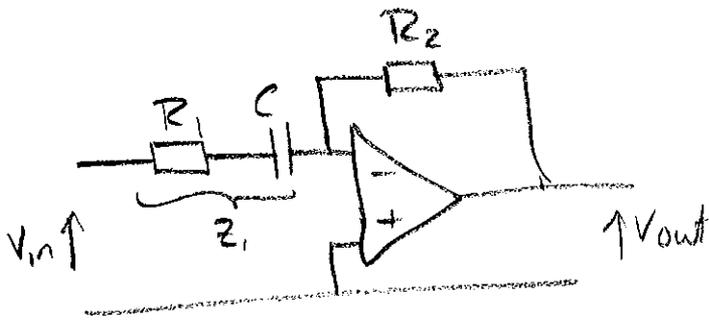
Instead

Small voltage $A_v = \frac{10+18+4}{10} = 3,2$

Large voltage $A_v = \frac{10+18}{10} = 2,8$

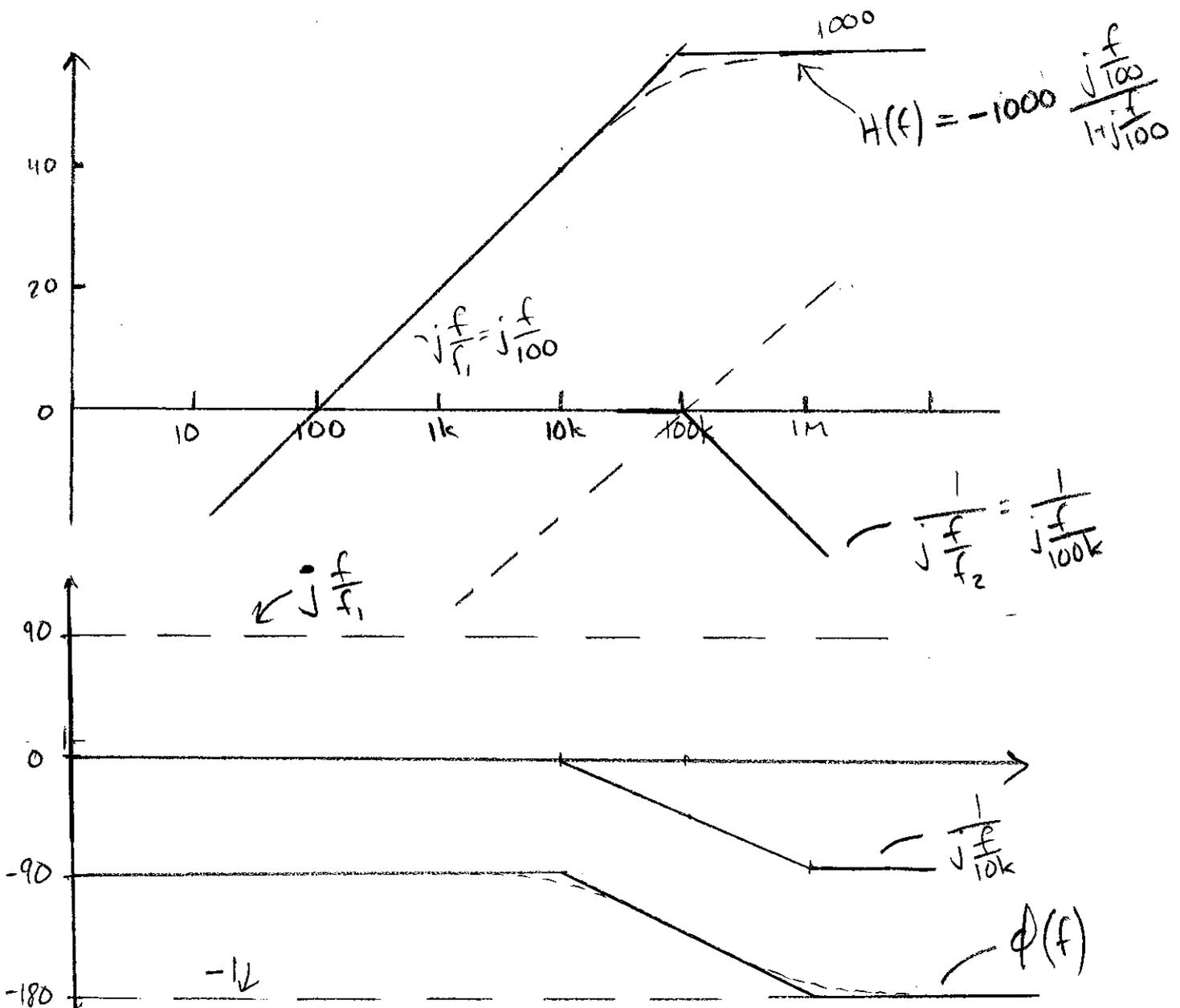


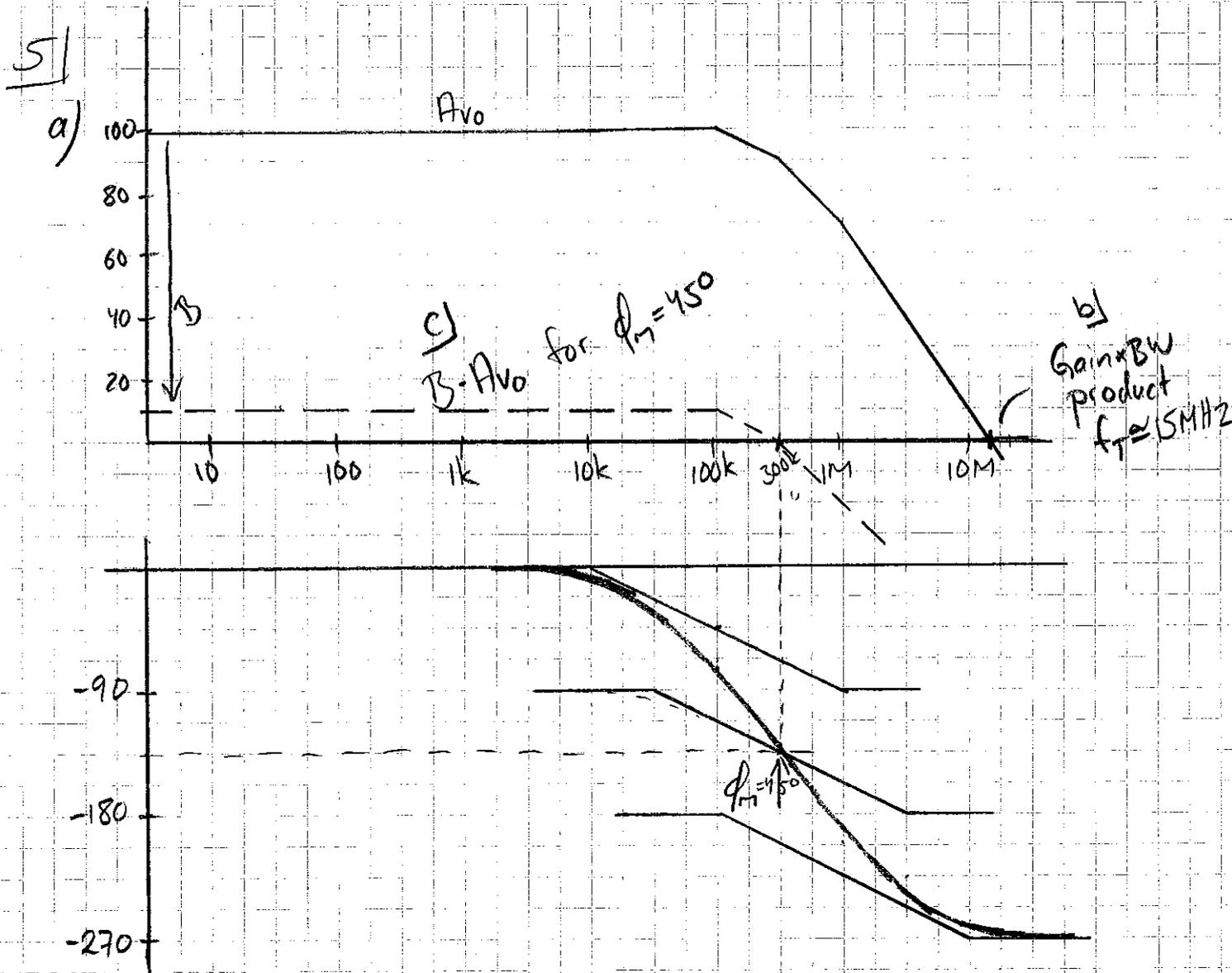
4



$$\frac{V_{out}}{V_{in}} = -\frac{Z_2}{Z_1} = \frac{R_2}{R_1 + \frac{1}{j\omega C}} = \frac{j\omega R_2 C}{1 + j\omega R_1 C} = -\frac{R_2}{R_1} \frac{j\omega R_1 C}{1 + j\omega R_1 C} = -\frac{R_2}{R_1} \frac{j \frac{f}{f_1}}{1 + j \frac{f}{f_1}}$$

where $f_1 = \frac{1}{2\pi R_1 C} = \frac{1}{2\pi \cdot 40 \cdot 40n} = 100k$



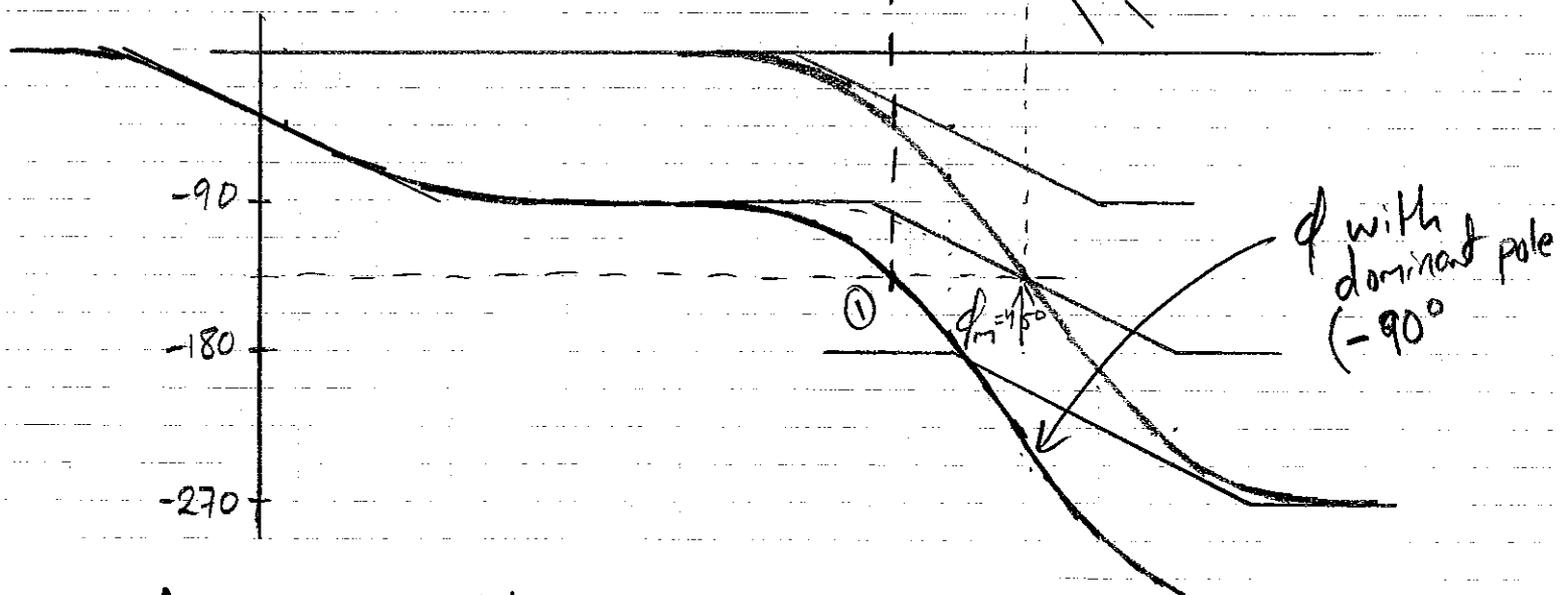
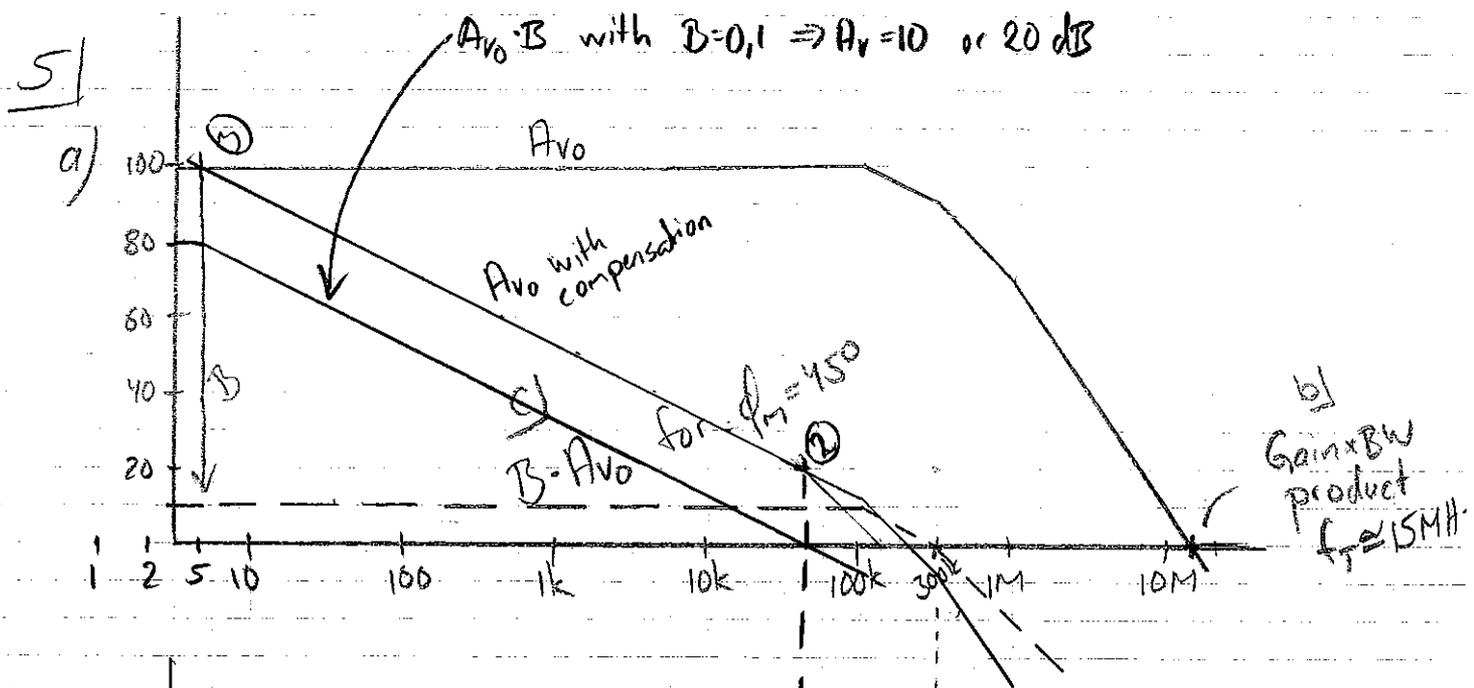


b) Gain \times BW product = $f_T = 15 \text{ MHz}$

c) Feed-back factor must be $< -90 \text{ dB}$ for 45° phase margin.

$$\Rightarrow A_v = \frac{1}{B} = 90 \text{ dB}$$

Amplifier stable for gain $> A_v = 90 \text{ dB}$



d) For stability @ 20 dB gain we add one pole that gives $A_{v0} = 20$ dB where $\phi = -135^\circ$
 (1) - (2) - (3)

$$f_{\text{pole}} \approx 5 \text{ Hz}$$